Department of Computer Engineering & Informatics

Laboratory for Signal Processing and Communications



Wireless and Mobile Communications

Key Technologies:

Efficient Modulation Techniques





- Modulation: Introductory concepts
- Main modulation techniques
 - Geometric representation
 - Modulation categories
 - Performance metrics
- Adaptive Modulation
- Hierarchical or layered modulation



Introductory concepts (1/5)







 Modulation: The process of "transforming" the information of a source so as to become suitable for transmission.

In the general case, it involves the transmission of information in a passband channel. The baseband signal is the "modulating" signal while the resulting passpand signal is the modulated one.

• Demodulation: The process of extracting the baseband signal from the bandwidth signal.

• The development of modulation techniques for mobile systems has been a subject of intensive research in the past 2-3 decades.





• In the mobile communication environment, some important features must be taken into account, such as:

- The multi-path phenomenon and the problems it causes (mainly due to frequency selective and/or fast fading).
- The problem of interference from other users.
- Non-linear distortions.
- Complexity / Implementation cost.
- The need for efficient use of energy (power) efficiency.
- The urgent, very often, need for spectral efficiency.
- The following criteria are **conflicting** and cannot be met at the same time

• Power efficiency: Rate of transmitted information per Watt of power (other indicators: CNR, E_s/N_0)

• Spectral efficiency: Rate of transmitted information per Hz of the available bandwidth (bits/sec/Hz).





Advantages of digital modulation and transmission:

- Resistance to noise and channel imperfections.
- Flexibility in multiplexing different forms of information.
- Integration of debugging, balancing, source coding procedures.
- Flexibility in terms of hardware and software implementation.

Basic techniques:

- Amplitude Shift Keying (ASK)
- Frequency Shift Keying (FSK)
- Phase Shift Keying (PSK)
- Amplitude and Phase Shift (QAM)





Introductory concepts (5/5)



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Digital modulation: setting up a suitable one-to-one correspondence between symbols s_i and waveforms $s_i(t)$, where i = 1, 2, ..., M.



• The different modulation techniques differ in the way the set of waveform signals $S = \{s_1(t), s_2(t), \dots, s_M(t)\}$ is defined.

Geometric representation:

■ Signals (and respective symbols) as points in a vector space.

$$s_i(t) = \sum_{m=1}^N s_{im} \phi_m(t)$$

• where $\phi_m(t)$, m = 1, 2, ..., N, the N-D space basis functions.

• $s_{im}, m = 1, 2, ..., N$, the coordinates of the signal $s_i(t)$ (i.e., the elements of the vector representing the signal $s_i(t)$)





• The constellation diagram for N = 2.

• Since $s_i(t) = s_{i1}\phi_1(t) + s_{i2}\phi_2(t)$, the vector representing the $s_i(t)$ is $\mathbf{s}_i = (s_{i1}, s_{i2})$.



• The constellation diagram provides useful information about the modulation system represented.

- Frequency efficiency increases with the density of the symbols.
- The required bandwidth increases with dimension N.
- The minimum distance between symbols determines the error probability.
- How energy efficiency could be improved ?





■ *M*-ary Quadrature Amplitude Modulation (*M*-QAM)

- The *M* symbols correspond to the signals $s_i(t) = A_i g_T(t) \cos(2\pi f_c t + \theta_i), i = 1, 2, ..., M$
- An alternative form of QAM signals is the following $s_i(t) = A_{ic}g_T(t)\cos(2\pi f_c t) + A_{is}g_T(t)\sin(2\pi f_c t), i = 1, 2, ..., M$





Quadrature Amplitude Modulation (cont.)









• Depending on whether the modulation technique requires synchronization between transmitter and receiver. For example, the FSK has two versions that differ in the way the receiver is implemented.

- Coherent FSK
 - Received signal: $r(t) = f(\cos(2\pi f_c t + 2\pi i\Delta f t + \phi_i))$
 - Requires knowledge of the ϕ_i
- Incoherent FSK
 - Received signal:

 $r(t) = f(\cos(\phi_i)\cos(2\pi f_i t) + \sin(\phi_i)\sin(2\pi f_i t))$





Block diagram of 2-FSK receiver in the incoherent case







Spectral efficient techniques:

Performance of the so-called
antipode techniques, such as *M*-PAM,
M-QAM, *M*-PSK







Power efficient techniques:

Performance of the so-called orthogonal techniques, such as
M-PPM and M-FSK







• The aforementioned modulation systems do not take into account the current state of the channel.

- Example of a non-adaptive modulation technique:
 - M-QAM, M = 4, 16, 64.
 - Suppose we want $P_M \le 10^{-4}$.
 - How can we achieve this?
 - If we additionally want to improve spectral efficiency?
 - If at the same time we wanted to maximize coverage?
 - What problems might arise?







- Drawbacks of non-adaptive modulation techniques:
 - □ The wireless channels are time varying, which implies that the instantaneous *SNR* is also varying, as well as the resulting probability of error
 - □ The instantaneous *SNR* is a function of the transmission conditions: y(t) = a(t)s(t) + w(t)
 - The channel gain a(t) depends on transmitter-receiver distance, multipath, weather conditions, etc.
 - □ If the selection of a (non-adaptive) modulation technique was simply based on the expected values of *SNR*, then we could not ensure a specified maximum probability of error.
 - □ For example, a minimum instantaneous *SNR* might also be required in addition to setting a maximum distance.
 - In the above cases, the choice of the modulation technique would be made so as to satisfy "worst case" conditions, regardless of how often these cases occur.





■ In adaptive modulation, the system is able to continuously adapt to the varying channel condition.

■ Adaptive techniques have been adopted by various standards such as IEEE 802.11 (Wi-Fi), IEEE 802.16 (Wi-Max), LTE advanced and 5G.

■ In order to apply adaptive techniques, we need to have a mechanism that estimates in real-time the channel parameters (e.g., the current value of *SNR*)

■ This information is fed back to the transmitter, which, in turn, makes the necessary parameter changes of the adopted modulation technique.



In what cases is "channel forecasting" needed?





■ In hierarchical (or layered) modulation, information is modulated at two levels:

- 1° High Priority Level (HP) where a robust but low-rate flow is generated (e.g., QPSK, why?).
- 2° Low Priority Level (LP) where a less robust but high-rate flow is generated (e.g., 64-QAM).

■ The two levels are formed into a common signal which is transmitted to the receiver.

■ This type of modulation is used, for example, in digital terrestrial video transmission (DVB-T).

• A receiver with poor reception conditions is limited to decoding the information present at the first level,

• If the channel condition allow the second level is also decoded.

It can also support easy system upgrades while maintaining "backwards" compatibility.





Example: 64-QAM (by design each symbol corresponds to 6 bits)

First layer: 2 bits (QPSK)

Second layer: 4 bits







- We will focus on the case of a frequently flat channel that changes slowly.
- The signal received at the receiver can be written as

$$r(t) = a(t)e^{-\theta(t)}s(t) + n(t), 0 \le t \le T$$

- $\alpha(t), \theta(t)$ denote the effect of the channel on the amplitude and phase of the signal s(t), T the symbol period and n(t) is white Gaussian noise.
- To calculate the error probability of a modulation technique we rely on the corresponding error probability when the channel is AWGN.

$$r(t) = s(t) + n(t), 0 \le t \le T$$

• Let this error probability be

$$P_{AWGN} = f(X)$$

- Where X is the signal-to-noise ratio (usually E_b/N_0).
- In case the wireless channel X is a function of a(t).





• To calculate the error probability $P_e = g(X)$ in the case of wireless channels, it is assumed, in principle, that the channel gain is constant.

- For a given channel gain α , the probability of error is $P_e(X(\alpha)) = P_{AWGN}(\alpha^2 E_b/N_0)$
- The error probability P_e is calculated as

$$P_e = \int_0^{+\infty} P_e(X(a)) p(X(a)) dX(a)$$

The probability density function (pdf) p(X(a)) is derived from the pdf of the channel gain α .

• For example, if the gain α is described by Rayleigh pdf, then

$$p(X(a)) = \frac{1}{\Gamma} e^{-\frac{X(a)}{\Gamma}}$$
, όπου $\Gamma = \frac{E_b}{N_0} \overline{a^2}$





- The GMSK as an example.
 - When the channel is AWGN

$$P_{AWGN} = Q \left\{ \sqrt{\frac{2\gamma E_b}{N_0}} \right\},\,$$

• $\gamma = 0,68$ and $\gamma = 0,85$ when $B_0T_b = 0,25$ and $B_0T_b = \infty$, respectively.

• When the channel is wireless (with the previous views) then

$$P_e = \frac{1}{2} \left(1 - \sqrt{\frac{\gamma \Gamma}{\gamma \Gamma + 1}} \right)$$

■ The case of BPSK.

- AWGN: $P_{AWGN} = Q\{\sqrt{2E_b/N_o}\}$
- Wireless channel: $P_e = \frac{1}{2} \left[1 \sqrt{\Gamma/(1+\Gamma)} \right]$

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• Comparison of BPSK when the transmission is done through the two different channel models considering $\overline{a^2} = 1$.

